

# Exam Winter Semester 2022

## Student Group

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### Exercise E4 Resistance of a Wire by Resistivity (written test, approx. 6 % of a 60-minute written test, WS2022)

A heating element made of solid nichrome wire with a diameter of  $d = 1.80 \text{ mm}$  is used for electric power dissipation (= heat flow) of  $P = 40 \text{ W}$  is necessary.

Determine the current  $I$  linked to the operating voltage for heating elements.

The Nichrome wire has a resistivity of  $\rho = 1.10 \cdot 10^{-6} \text{ } \Omega \cdot \text{m}$ .

The heating element is  $l = 3 \text{ m}$  long and has a diameter of  $d = 3.57 \text{ mm}$ .

∴ Calculate the resistance  $R$  of the heating element.

Solution

$$\begin{aligned} P &= U \cdot I = R \cdot I^2 \quad \rightarrow \quad I = \\ &= \sqrt{\frac{P}{R}} = \sqrt{\frac{40 \text{ W}}{0.33 \text{ } \Omega}} \end{aligned}$$

$$\begin{aligned} R &= \rho \cdot \frac{l}{A} \quad \text{with } A = r^2 \cdot \pi = \\ &= \frac{1}{4} d^2 \cdot \pi \quad \text{and } R = \rho \cdot \frac{l}{\frac{1}{4} d^2 \cdot \pi} \quad \text{and } R = \\ &= 1.10 \cdot 10^{-6} \text{ } \Omega \cdot \text{m} \cdot \frac{4 \cdot 3 \text{ m}}{(3.57 \cdot 10^{-3} \text{ m})^2 \cdot \pi} \end{aligned}$$

### Exercise E1 Resistance of a Wire by Resistivity (written test, approx. 6 % of a 60-minute written test, WS2022)

A heating element made of solid nichrome wire with a diameter of  $d = 1.80 \text{ mm}$  is used for electric power dissipation (= heat flow) of  $P = 40 \text{ W}$  is necessary.

Determine the current  $I$  linked to the operating voltage for heating elements.

The Nichrome wire has a resistivity of  $\rho = 1.10 \cdot 10^{-6} \text{ } \Omega \cdot \text{m}$ .

The heating element is  $l = 3 \text{ m}$  long and has a diameter of  $d = 3.57 \text{ mm}$ .

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Solution

$$\begin{aligned} P &= U \cdot I = R \cdot I^2 \quad \rightarrow \quad I = \\ &= \sqrt{\frac{P}{R}} = \sqrt{\frac{40 \text{ W}}{0.33 \text{ } \Omega}} \end{aligned}$$

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### Exercise E1 Temperature-dependent Resistance

**(written test, approx. 6 % of a 60-minute written test, WS2022)**

2. The diagram shows a circuit with a resistor and a diode. The diode has a resistance of  $10 \text{ k}\Omega$  at  $25^\circ\text{C}$ . Its temperature coefficients are:  $\alpha = 0.01 \text{ K}^{-1}$  and  $\beta = 71 \cdot 10^{-6} \text{ K}^{-2}$ .

Result: The temperature inside the refrigeration system can reach down to  $-40^\circ\text{C}$ .

Calculate the resistance of the thermal resistor at  $-40^\circ\text{C}$ .

$$R = 10 \text{ k}\Omega \cdot (1 + 0.01 \cdot (-40 - 25) + 71 \cdot 10^{-6} \cdot (-40 - 25)^2)$$

The power transfer resistor  $P$  is a part of the circuit and generates heat. Therefore, a solution is to increase the heat flow up the refrigeration system.

Therefore, with constant  $U$  and increasing  $R$  the power decreases. Ten times more resistance decreases the heat flow to one-tenth.

$$R = R_0 \cdot (1 + \alpha \cdot \Delta T + \beta \cdot \Delta T^2)$$

with  $\Delta T = T_{\text{end}} - T_{\text{start}}$

$$R = 10 \text{ k}\Omega \cdot (1 + 0.01 \cdot (-40^\circ\text{C} - 25^\circ\text{C}) + 71 \cdot 10^{-6} \cdot (-40^\circ\text{C} - 25^\circ\text{C})^2)$$

**Exercise E2 Temperature-dependent Resistance (written test, approx. 6 % of a 60-minute written test, WS2022)**

2. The diagram shows a circuit with a resistor and a diode. The diode has a resistance of  $10 \text{ k}\Omega$  at  $25^\circ\text{C}$ . Its temperature coefficients are:  $\alpha = 0.01 \text{ K}^{-1}$  and  $\beta = 71 \cdot 10^{-6} \text{ K}^{-2}$ .

Result: The temperature inside the refrigeration system can reach down to  $-40^\circ\text{C}$ .

Calculate the resistance of the thermal resistor at  $-40^\circ\text{C}$ .

$$R = 10 \text{ k}\Omega \cdot (1 + 0.01 \cdot (-40 - 25) + 71 \cdot 10^{-6} \cdot (-40 - 25)^2)$$

The power transfer resistor  $P$  is a part of the circuit and generates heat. Therefore, a solution is to increase the heat flow up the refrigeration system.

Therefore, with constant  $U$  and increasing  $R$  the power decreases. Ten times more resistance decreases the heat flow to one-tenth.

$$R = R_0 \cdot (1 + \alpha \cdot \Delta T + \beta \cdot \Delta T^2)$$

with  $\Delta T = T_{\text{end}} - T_{\text{start}}$

$$R = 10 \text{ k}\Omega \cdot (1 + 0.01 \cdot (-40^\circ\text{C} - 25^\circ\text{C}) + 71 \cdot 10^{-6} \cdot (-40^\circ\text{C} - 25^\circ\text{C})^2)$$

**Exercise E6 Pure Resistor Network Simplification**  
**(written test, approx. 13 % of a 60-minute written test, WS2022)**

The following shall be solved at once, the given resistances and the voltage  
 result given.  $R_2 = 400 \Omega$ ,  $R_3 = 100 \Omega$ ,  $R_4 = 100 \Omega$ ,  $R_5 = 100 \Omega$ ,  $R_6 = 100 \Omega$ ,  $R_7 = 100 \Omega$ ,  $R_8 = 100 \Omega$ ,  $R_9 = 100 \Omega$ ,  $R_{10} = 100 \Omega$ ,  $R_{11} = 100 \Omega$ ,  $R_{12} = 100 \Omega$ ,  $R_{13} = 100 \Omega$ ,  $R_{14} = 100 \Omega$ ,  $R_{15} = 100 \Omega$ ,  $R_{16} = 100 \Omega$ ,  $R_{17} = 100 \Omega$ ,  $R_{18} = 100 \Omega$ ,  $R_{19} = 100 \Omega$ ,  $R_{20} = 100 \Omega$ ,  $R_{21} = 100 \Omega$ ,  $R_{22} = 100 \Omega$ ,  $R_{23} = 100 \Omega$ ,  $R_{24} = 100 \Omega$ ,  $R_{25} = 100 \Omega$ ,  $R_{26} = 100 \Omega$ ,  $R_{27} = 100 \Omega$ ,  $R_{28} = 100 \Omega$ ,  $R_{29} = 100 \Omega$ ,  $R_{30} = 100 \Omega$ ,  $R_{31} = 100 \Omega$ ,  $R_{32} = 100 \Omega$ ,  $R_{33} = 100 \Omega$ ,  $R_{34} = 100 \Omega$ ,  $R_{35} = 100 \Omega$ ,  $R_{36} = 100 \Omega$ ,  $R_{37} = 100 \Omega$ ,  $R_{38} = 100 \Omega$ ,  $R_{39} = 100 \Omega$ ,  $R_{40} = 100 \Omega$ ,  $R_{41} = 100 \Omega$ ,  $R_{42} = 100 \Omega$ ,  $R_{43} = 100 \Omega$ ,  $R_{44} = 100 \Omega$ ,  $R_{45} = 100 \Omega$ ,  $R_{46} = 100 \Omega$ ,  $R_{47} = 100 \Omega$ ,  $R_{48} = 100 \Omega$ ,  $R_{49} = 100 \Omega$ ,  $R_{50} = 100 \Omega$ ,  $R_{51} = 100 \Omega$ ,  $R_{52} = 100 \Omega$ ,  $R_{53} = 100 \Omega$ ,  $R_{54} = 100 \Omega$ ,  $R_{55} = 100 \Omega$ ,  $R_{56} = 100 \Omega$ ,  $R_{57} = 100 \Omega$ ,  $R_{58} = 100 \Omega$ ,  $R_{59} = 100 \Omega$ ,  $R_{60} = 100 \Omega$ ,  $R_{61} = 100 \Omega$ ,  $R_{62} = 100 \Omega$ ,  $R_{63} = 100 \Omega$ ,  $R_{64} = 100 \Omega$ ,  $R_{65} = 100 \Omega$ ,  $R_{66} = 100 \Omega$ ,  $R_{67} = 100 \Omega$ ,  $R_{68} = 100 \Omega$ ,  $R_{69} = 100 \Omega$ ,  $R_{70} = 100 \Omega$ ,  $R_{71} = 100 \Omega$ ,  $R_{72} = 100 \Omega$ ,  $R_{73} = 100 \Omega$ ,  $R_{74} = 100 \Omega$ ,  $R_{75} = 100 \Omega$ ,  $R_{76} = 100 \Omega$ ,  $R_{77} = 100 \Omega$ ,  $R_{78} = 100 \Omega$ ,  $R_{79} = 100 \Omega$ ,  $R_{80} = 100 \Omega$ ,  $R_{81} = 100 \Omega$ ,  $R_{82} = 100 \Omega$ ,  $R_{83} = 100 \Omega$ ,  $R_{84} = 100 \Omega$ ,  $R_{85} = 100 \Omega$ ,  $R_{86} = 100 \Omega$ ,  $R_{87} = 100 \Omega$ ,  $R_{88} = 100 \Omega$ ,  $R_{89} = 100 \Omega$ ,  $R_{90} = 100 \Omega$ ,  $R_{91} = 100 \Omega$ ,  $R_{92} = 100 \Omega$ ,  $R_{93} = 100 \Omega$ ,  $R_{94} = 100 \Omega$ ,  $R_{95} = 100 \Omega$ ,  $R_{96} = 100 \Omega$ ,  $R_{97} = 100 \Omega$ ,  $R_{98} = 100 \Omega$ ,  $R_{99} = 100 \Omega$ ,  $R_{100} = 100 \Omega$ .

Solution

$$R_{\text{eq}} = 132.8 \Omega$$

Now a wye-delta transformation is necessary.

Since  $R_2 = R_3$  and based on the equations for the transformation, the transformed  $R_Y$  is given as:

$$R_Y = \frac{R_2 \cdot R_3}{R_2 + R_3} = \frac{100 \cdot 100}{100 + 100} = 50 \Omega$$

The equivalent resistor is given by a parallel configuration of resistors in series:

$$R_{\text{eq}} = R_Y + (R_Y + R_1 + R_1) \parallel (R_Y + R_2) \parallel (R_Y + R_2) \parallel (R_Y + R_2) \parallel (R_Y + R_2)$$

The switch shall now be open. Calculate the equivalent resistance  $R_{\text{eq}}$  between  $A$  and  $B$ .

Solution



The equivalent resistor is given by a parallel configuration of resistors in series:

$$R_{\text{eq}} = (R_2 + R_1 + R_1) \parallel (R_2 + R_2) \parallel R_{\text{eq}} = (100 \sim\Omega + 200 \sim\Omega + 200 \sim\Omega) \parallel (100 \sim\Omega + 100 \sim\Omega) \parallel R_{\text{eq}} = \{500 \sim\Omega\} \parallel (200 \sim\Omega) \parallel R_{\text{eq}} = \{500 \sim\Omega \cdot 200 \sim\Omega\} \over {500 \sim\Omega + 200 \sim\Omega}$$

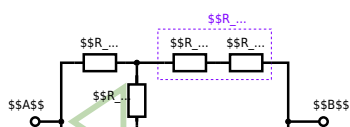
**Exercise E3 Pure Resistor Network Simplification (written test, approx. 13 % of a 60-minute written test, WS2022)**

The following shall be solved with  $R_1 = 200 \Omega$ ,  $R_2 = R_3 = 100 \Omega$  and the switch  $B$  is given.

Solution

$$R_{\text{eq}} = 132.8 \Omega$$

Now a wye-delta transformation is necessary.

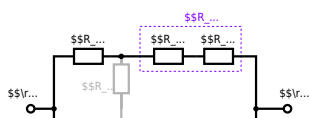


Since  $R_2=R_3$  and based on the equations for the transformation, the transformed  $R_Y$  is given as: 
$$R_Y = \frac{R_2 \cdot R_2}{R_2 + R_2 + R_2} = \frac{(100 \Omega)^2}{3 \cdot 100 \Omega} = \frac{1}{3} \cdot 100 \Omega = 33.33 \Omega$$

The equivalent resistor is given by a parallel configuration of resistors in series: 
$$R_{eq} = R_Y + (R_Y + R_1 + R_1) \parallel (R_Y + R_2) \parallel R_{eq} = 33.33 \Omega + (33.33 \Omega + 400 \Omega) \parallel (33.33 \Omega + 100 \Omega)$$

1. The switch shall now be open. Calculate the equivalent resistance  $R_{eq}$  between A and B.

Solution



The equivalent resistor is given by a parallel configuration of resistors in series:

$$R_{\text{eq}} = (R_2 + R_1 + R_{-1}) \parallel (R_2 + R_2) \parallel R_{\text{eq}} = (100 \Omega + 200 \Omega + 200 \Omega) \parallel (100 \Omega + 100 \Omega) \parallel R_{\text{eq}} = (500 \Omega) \parallel (200 \Omega) \parallel R_{\text{eq}} = \frac{500 \Omega \cdot 200 \Omega}{500 \Omega + 200 \Omega} \parallel$$

**Exercise E1 Equivalent linear Source  
(written test, approx. 14 % of a 60-minute written test, WS2022)**

The circuit in the following has to be simplified.  
Result

$$U_{\text{s}} = U_{\text{AB}} = 4.5 \text{ V} \quad R_{\text{i}} = R_{\text{AB}} = 6 \Omega$$



Calculate the internal resistance  $R_i$  and the source voltage  $U_s$  of an equivalent linear voltage source on the connectors  $A$  and  $B$ .  $R_1=5.0 \Omega$ ,  $U_2=6.0 \text{ V}$ ,  $R_3= 10 \Omega$ ,  $I_4=4.2 \text{ A}$ ,  $R_5=10 \Omega$ ,  $R_6=7.5 \Omega$ ,  $R_7=15 \Omega$  Use equivalent sources in order to simplify the circuit!

Solution

The best thing is to re-think the wiring like rubber bands and adjust them:



The linear voltage source of  $U_2$  and  $R_1$  can be transformed into a current source  $I_2 = \frac{U_2}{R_1}$  and  $R_1$ :



Now a lot of them can be combined. The resistors  $R_1$ ,  $R_3$ ,  $R_5$  are in parallel, like also  $I_2$  and  $I_4$ :

$$R_{135} = R_1 || R_3 || R_5$$

$$I_{24} = I_2 - I_4 = \frac{U_{24}}{R_1} - I_4$$

The resulting circuit can again be transformed:



Here, the  $U_{24}$  is calculated by  $I_{24}$  as the following:

$$U_{24}$$

$$U_{24} = U_2 \cdot \frac{R_7}{R_6 + R_7 + R_1 || R_3 || R_5} - I_4 \cdot R_1 || R_3 || R_5$$

On the right side of the last circuit, there is a voltage divider given by  $R_{135}$ ,  $R_6$ , and  $R_7$ .

Therefore the voltage between  $A$  and  $B$  is given as:

$$U_{AB} = U_{24} \cdot \frac{R_7}{R_6 + R_7 + R_1 || R_3 || R_5} - I_4 \cdot \frac{R_7 \cdot R_1 || R_3 || R_5}{R_6 + R_7 + R_1 || R_3 || R_5}$$

For the internal resistance  $R_i$  the ideal voltage source is substituted by its resistance ( $=0\Omega$ , so a short-circuit):

$$R_{AB} = R_7 || (R_6 + R_1 || R_3 || R_5)$$

with  $R_1 || R_3 || R_5 = 5\Omega || 10\Omega || 10\Omega = 5\Omega || 5\Omega = 2.5\Omega$ :

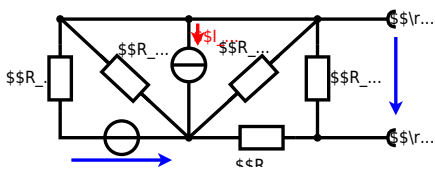
$$U_{AB} = \frac{6.0\text{V}}{5.0\Omega} - 4.2\Omega \cdot \frac{15\Omega \cdot 2.5\Omega}{7.5\Omega + 15\Omega + 2.5\Omega}$$

$$R_{AB} = 15\Omega || (7.5\Omega + 2.5\Omega)$$

### Exercise E3 Equivalent linear Source (written test, approx. 14 % of a 60-minute written test, WS2022)

The circuit in the following has to be simplified.  
Result

$$U_s = U_{AB} = 4.5\text{V} \quad R_i = R_{AB} = 6\Omega$$



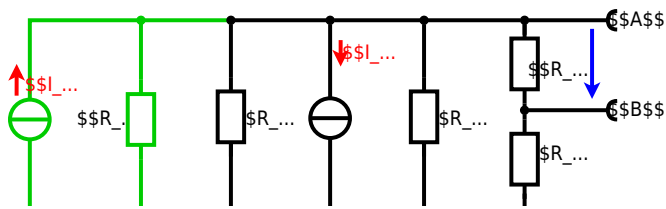
Calculate the internal resistance  $R_{in}$  and the source voltage  $U_{s}$  of an equivalent linear voltage source on the connectors  $A$  and  $B$ .  
 $R_1=5.0 \Omega$ ,  $U_2=6.0 \text{ V}$ ,  $R_3= 10 \Omega$ ,  $I_4=4.2 \text{ A}$ ,  
 $R_5=10 \Omega$ ,  $R_6=7.5 \Omega$ ,  $R_7=15 \Omega$  Use equivalent sources in order to simplify the circuit!

Solution

The best thing is to re-think the wiring like rubber bands and adjust them:



The linear voltage source of  $U_2$  and  $R_1$  can be transformed into a current source  $I_2 = \frac{U_2}{R_1}$  and  $R_1$ :



Now a lot of them can be combined. The resistors  $R_1$ ,  $R_3$ ,  $R_5$  are in parallel, like also  $I_2$  and  $I_4$ :

$$R_{135} = R_1 || R_3 || R_5$$

$$I_{24} = I_2 - I_4 = \frac{U_2}{R_1} - I_4$$

The resulting circuit can again be transformed:



Here, the  $U_{24}$  is calculated by  $I_{24}$  as the following:

$$U_{24} = I_{24} \cdot R_{67}$$

$$U_{AB} = U_{24} \cdot \frac{R_7}{R_6 + R_7 + R_1 || R_3 || R_5} - I_4 \cdot R_1 || R_3 || R_5$$

On the right side of the last circuit, there is a voltage divider given by  $R_{135}$ ,  $R_6$ , and  $R_7$ .

Therefore the voltage between  $A$  and  $B$  is given as:

$$U_{AB} = U_{24} \cdot \frac{R_7}{R_6 + R_7 + R_1 || R_3 || R_5} - (U_2 \cdot \frac{1}{R_1} - I_4) \cdot R_1 || R_3 || R_5$$

For the internal resistance  $R_i$  the ideal voltage source is substituted by its resistance ( $=0 \Omega$ , so a short-circuit):

$$R_{AB} = R_7 || (R_6 + R_1 || R_3 || R_5)$$

with  $R_1 || R_3 || R_5 = 5 \Omega || 10 \Omega || 10 \Omega = 5 \Omega || 5 \Omega = 2.5 \Omega$ :

$$U_{AB} = \frac{6.0 \text{ V}}{5.0 \Omega} - 4.2 \Omega \cdot \frac{15 \Omega \cdot 2.5 \Omega}{7.5 \Omega + 15 \Omega + 2.5 \Omega}$$

$$R_{AB} = 15 \Omega || (7.5 \Omega + 2.5 \Omega)$$

**Exercise E5 Charging Capacitors**  
**(written test, approx. 16 % of a 60-minute written test, WS2022)**

The circuit below is a fully discharging RC circuit. The capacitor has a capacitance of  $C = 2 \mu\text{F}$  and is initially at  $U_C = 0 \text{ V}$ . The voltage across the capacitor is again  $0 \text{ V}$  at the moment  $t_0 = 0 \text{ s}$  when the switch  $S_1$  is closed. Calculate the voltage  $u_c(t_2)$  across the capacitor at  $t_2 = 1 \text{ ms}$  after closing the switch.

**Solution:** To solve this, first create an equivalent linear voltage source from  $U$ ,  $R_1$ , and  $R_2$ .

$$U_{eq} = U \cdot \frac{R_2}{R_1 + R_2} = 12 \text{ V} \cdot \frac{2 \Omega}{2 \Omega + 2 \Omega} = 6 \text{ V}$$

On an alternative view, one can try to create an equivalent linear voltage source again. Then, the internal resistance is given by substituting the ideal voltage source is again short-circuiting  $R_2$ .



The circuit contains a voltage source  $U=12\text{ V}$ , a switch  $S_1$ , a resistor of  $R_1=20\text{ }\Omega$  and a capacitor of  $C=100\text{ }\mu\text{F}$ . The switch  $S_2$  to an additional consumer  $R_2$  will be considered to be open for the first task. At the moment  $t_0=0\text{ s}$  the switch  $S_1$  is closed, the voltage across the capacitor is  $u_c(t_0)=0\text{ V}$ .



First do not consider the light bulb - it is not connected to the RC circuit. Calculate the point of time  $t_1$  when  $u_c(t_1)=0.5 \cdot U$ .

Solution



So, here only  $R_1$  and  $C$  gives the time constant:  $\tau = R_1 \cdot C$

The following formula describes the time course of  $u_c(t)$  which has to be  $u_c(t_1)=0.5 \cdot U$ :  

$$u_c(t) = U \cdot (1 - e^{-t/\tau}) = 0.5 \cdot U$$
 It has to be rearranged to  $(1 - e^{-t/\tau}) = 0.5$ .  
 An equivalent linear voltage source can be given with  $U_s$ ,  $R_1$  and  $R_B$  as seen in yellow:  

$$U_s = U \cdot \frac{R_B}{R_1 + R_B} = 1/2 \cdot U$$
 The internal resistance is given by substituting the ideal voltage source with its resistance ( $=0\text{ }\Omega$ , short-circuit).  

$$R_i = R_1 \parallel R_B = 10\text{ }\Omega$$

Therefore, the voltage of the equivalent linear voltage source is:  

$$u_c(t_2) = U_s \cdot (1 - e^{-t_2/(R_i \cdot C)}) = \frac{1}{2} \cdot U \cdot (1 - e^{-1\text{ ms}/(10\text{ }\Omega \cdot 100\text{ }\mu\text{F})})$$

**Exercise E4 Charging Capacitors**  
 (written test, approx. 16 % of a 60-minute written test, WS2022)

The circuit (as shown in the figure) consists of a DC voltage source  $U=6\text{ V}$ , a resistor  $R_1=20\text{ }\Omega$ , a capacitor  $C=20\text{ }\mu\text{F}$  and a light bulb  $R_B=20\text{ }\Omega$ . The switch  $S_1$  is open. The voltage across the capacitor is again  $0\text{ V}$  at the moment  $t_0=0\text{ s}$  when the switch  $S_1$  is closed. Calculate the voltage  $u_c(t_2)$  across the capacitor at  $t_2=1\text{ ms}$  after closing the switch.

**Solution**  
 To solve this, first create an equivalent linear voltage source from  $U$ ,  $R_1$ , and  $R_B$ .  

$$U_s = U \cdot \frac{R_B}{R_1 + R_B} = 3\text{ V}$$

$$R_i = R_1 \parallel R_B = 10\text{ }\Omega$$

Solution

The ideal voltage source is  $U = 12 \text{ V}$ . The internal resistance is given by substituting the ideal voltage source is again short-circuiting  $R_2$ .

On an alternative view, one can try to create an equivalent linear voltage source again. Then, the internal resistance is given by substituting the ideal voltage source is again short-circuiting  $R_2$ .



The circuit contains a voltage source  $U = 12 \text{ V}$ , a switch  $S_1$ , a resistor of  $R_1 = 20 \text{ }\Omega$  and a capacitor of  $C = 100 \text{ }\mu\text{F}$ .

The switch  $S_2$  to an additional consumer  $R_2$  will be considered to be open for the first asks. At the moment  $t_0 = 0$  the switch  $S_1$  is closed, the voltage across the capacitor is  $u_c(t_0) = 0$ .

First do not consider the light bulb - it is not connected to the RC circuit. Calculate the point of time  $t_1$  when  $u_c(t_1) = 0.5 \cdot U$ .



Solution

An equivalent linear voltage source can be given with  $U_s$ ,  $R_1$ , and  $R_B$  as seen in yellow.

Therefore, the voltage of the equivalent linear voltage source is:  $U_s = U \cdot \frac{R_B}{R_1 + R_B} = \frac{1}{2} \cdot U$ . The internal resistance is given by substituting the ideal voltage source with its resistance ( $R = 0 \text{ }\Omega$ , short-circuit).  $R_i = R_1 \parallel R_B = 10 \text{ }\Omega$

$$u_c(t_2) = U_s \cdot (1 - e^{-t_2/(R_i \cdot C)}) = \frac{1}{2} \cdot U \cdot (1 - e^{-t_2/(10 \text{ }\Omega \cdot 100 \text{ }\mu\text{F})})$$

So, here only  $R_1$  and  $C$  gives the time constant:  $\tau = R_1 \cdot C$

The following formula describes the time course of  $u_c(t)$  which has to be  $u_c(t_1) = 0.5 \cdot U$ :  $u_c(t) = U \cdot (1 - e^{-t/\tau}) = 0.5 \cdot U$ . It has to be rearranged to  $(1 - e^{-t/\tau}) = 0.5 \implies e^{-t/\tau} = 0.5 \implies -t/\tau = \ln(0.5) \implies t = \tau \cdot \ln(0.5)$

**Exercise E2 Analyzing complex Impedances**  
 (written test, approx. 14 % of a 60-minute written test, WS2022)

2. Given that the phasor voltage source  $\underline{U} = 50 \angle 0^\circ \text{ V}$  and the admittance  $\underline{Y} = 0.24 \text{ S}$  are both in the components. ( $\$R\$$  and  $\$X_L\$$ ) shall be given.

After analysis, the full width dimensioned complex impedance  $\underline{Z}$  shall be extracted and the magnitude  $|Z|$  in phase  $\varphi$  shall be given.  $\underline{Z} = (2 + j4) \Omega$

Solution  
 .. Calculation of physical values of the two components.  
 Solution  $\begin{aligned} R &= \frac{1}{0.24} = 4.17 \Omega \\ X_L &= \frac{1}{0.24} - R = 4.17 \Omega - 4.17 \Omega = 0 \Omega \end{aligned}$

Solution

$$\underline{I} = \frac{\underline{U}}{\underline{Z}} = \frac{50 \angle 0^\circ \text{ V}}{(2 + j4) \Omega} = 12.5 \angle -63.4^\circ \text{ A}$$

The current  $I$  has a magnitude of  $12.5 \text{ A}$  and a phase angle of  $-63.4^\circ$ . The voltage  $U$  is  $50 \text{ V}$  and the admittance  $Y$  is  $0.24 \text{ S}$ . The resulting impedance  $Z$  is  $(2 + j4) \Omega$ .

Therefore, the component  $R$  is  $4.17 \Omega$  and the component  $X_L$  is  $0 \Omega$ . The magnitude of the impedance  $Z$  is  $|Z| = \sqrt{2^2 + 4^2} = 4.47 \Omega$ .

The phase  $\varphi$  is  $\varphi = \arctan\left(\frac{4}{2}\right) = 63.4^\circ$ . The phase angle of the current  $I$  is  $-63.4^\circ$ .

With the complex part  $(2 + j4) \Omega$  the physical values  $R$  and  $X_L$  are  $R = 4.17 \Omega$  and  $X_L = 0 \Omega$ .

The phase  $\varphi$  shall be calculated as  $\varphi = \arctan\left(\frac{4}{2}\right) = 63.4^\circ$ .

The magnitude  $|Z|$  shall be calculated as  $|Z| = \sqrt{2^2 + 4^2} = 4.47 \Omega$ .

**Exercise E5 Analyzing complex Impedances**  
 (written test, approx. 14 % of a 60-minute written test, WS2022)

2. Given that the phasor voltage source  $\underline{U} = 50 \angle 0^\circ \text{ V}$  and the admittance  $\underline{Y} = 0.24 \text{ S}$  are both in the components. ( $\$R\$$  and  $\$X_L\$$ ) shall be given.

After analysis, the full width dimensioned complex impedance  $\underline{Z}$  shall be extracted and the magnitude  $|Z|$  in phase  $\varphi$  shall be given.  $\underline{Z} = (2 + j4) \Omega$

Solution  
 .. Calculation of physical values of the two components.  
 Solution  $\begin{aligned} R &= \frac{1}{0.24} = 4.17 \Omega \\ X_L &= \frac{1}{0.24} - R = 4.17 \Omega - 4.17 \Omega = 0 \Omega \end{aligned}$

Solution

$$\underline{I} = \frac{\underline{U}}{\underline{Z}} = \frac{50 \angle 0^\circ \text{ V}}{(2 + j4) \Omega} = 12.5 \angle -63.4^\circ \text{ A}$$

The current  $I$  has a magnitude of  $12.5 \text{ A}$  and a phase angle of  $-63.4^\circ$ . The voltage  $U$  is  $50 \text{ V}$  and the admittance  $Y$  is  $0.24 \text{ S}$ . The resulting impedance  $Z$  is  $(2 + j4) \Omega$ .

Therefore, the component  $R$  is  $4.17 \Omega$  and the component  $X_L$  is  $0 \Omega$ . The magnitude of the impedance  $Z$  is  $|Z| = \sqrt{2^2 + 4^2} = 4.47 \Omega$ .

The absolute value of the impedance is  $Z = \sqrt{R^2 + (X_L - X_C)^2}$  with  $R = 5 \Omega$ ,  $X_L = \omega L = 2\pi \cdot 4 \text{ MHz} \cdot 100 \text{ nH} = 2.51 \text{ m}\Omega$  and  $X_C = \frac{1}{\omega C} = \frac{1}{2\pi \cdot 4 \text{ MHz} \cdot 10 \text{ nF}} = 3.98 \text{ m}\Omega$ .  
 The phase  $\varphi$  is given by  $\varphi = \arctan\left(\frac{X_L - X_C}{R}\right) = \arctan\left(\frac{2.51 \text{ m}\Omega - 3.98 \text{ m}\Omega}{5 \Omega}\right) = -0.24 \text{ rad}$ .  
 With the complex part comes the physical value:  $I = \frac{U}{Z} = \frac{10 \text{ V}}{\sqrt{5^2 + (2.51 - 3.98)^2}} = 2.0 \text{ A}$ .  
 The phase  $\varphi$  is  $\varphi = -0.24 \text{ rad} = -13.7^\circ$ .

**Exercise E3 Impedances at different Frequencies**  
 (written test, approx. 18 % of a 60-minute written test, WS2022)

At a series circuit with a resistor  $R_1 = 1 \text{ k}\Omega$ , a capacitor  $C_1 = 40 \text{ nF}$  and an inductor  $L_1 = 4.7 \text{ }\mu\text{H}$  in AC with a voltage  $U = 10 \text{ V}$  and a frequency  $f = 4 \text{ MHz}$ .  
 Result:  $Z = 1.00 \text{ }\Omega$ ,  $I = 10 \text{ A}$ ,  $\varphi = 0^\circ$ .  
 A resistor  $R_1$  shall have the same absolute value of the impedance as a capacitor  $C_1 = 40 \text{ nF}$  at  $f_1 = 4 \text{ MHz}$ .

Solution  
 Solution  $R_1 = 1.00 \Omega$   
 Solution  $R_2 = 10.0 \Omega$

A series circuit means that the current is constant on every component.  
 The equivalent impedance for  $R$  and  $L$  combined is given by  $Z = \sqrt{R^2 + X_L^2}$ .  
 Parallel circuit means that the voltage is the same on  $R_2$  and  $C_2$ .  
 $Z = \frac{1}{\frac{1}{R_2} + \frac{1}{X_C}} = \frac{R_2 X_C}{R_2 + X_C}$ . Since  $X_C$  is perpendicular to  $R_2$ , this can be simplified to  $Z = \frac{R_2 X_C}{\sqrt{R_2^2 + X_C^2}}$ .  
 $X_C$  is perpendicular to  $X_L$  (It has to, since  $R_3$  is perpendicular to  $X_L$ ).  
 Therefore, the resulting current of the parallel circuit is given as:  
 $I = \frac{U}{Z} = \frac{U \sqrt{R_2^2 + X_C^2}}{R_2 X_C}$   
 This can be rearranged to get  $R_2 = \frac{U \sqrt{R_2^2 + X_C^2}}{I X_C}$   
 $R_2^2 = \frac{U^2 (R_2^2 + X_C^2)}{I^2 X_C^2}$   
 $R_2^2 (I^2 X_C^2 - U^2) = U^2 X_C^2$   
 $R_2 = \frac{U X_C}{\sqrt{I^2 X_C^2 - U^2}}$   
 Back to the first formula:  $R_3 \cdot I = X_C \cdot I$   
 $R_3 = \frac{X_C}{I} = \frac{1}{2\pi \cdot 4 \text{ MHz} \cdot 40 \text{ nF}} = 10.0 \text{ }\Omega$

**Exercise E6 Impedances at different Frequencies**  
 (written test, approx. 18 % of a 60-minute written test, WS2022)

**Resistor**  $R_1$  shall have the same absolute value of the impedance as a capacitor  $C_1 = 40 \text{ nF}$  at  $f_1 = 4 \text{ MHz}$ .

**Solution**

$$R_1 = 1.00 \text{ } \Omega$$

$$R_2 = 10.0 \text{ } \Omega$$

A series circuit means that the current is constant on every component.

The equivalent impedance for  $R$  and  $L$  combined is given by

$$Z_{RL} = R + j\omega L$$

Parallel circuit means that the voltage is the same on  $R_2$  and  $C_1$

$$Z_{RC} = \frac{R_2 \cdot (-j/\omega C_1)}{R_2 - j/\omega C_1}$$

Since  $Z_{RC}$  is perpendicular to  $Z_{RL}$ , the resulting current of the parallel circuit is given as:

$$I_{3R} = I_{2R} + I_{3C}$$

$$I_{3R} = \frac{U}{R_2} + \frac{U}{-j\omega C_1}$$

$$I_{3R} = \frac{U}{R_2} \left( 1 - j\omega R_2 C_1 \right)$$

$$I_{3R} = \frac{3.0 \text{ V}}{10 \text{ } \Omega} \left( 1 - j \cdot 2\pi \cdot 4 \cdot 10^6 \cdot 40 \cdot 10^{-9} \right)$$

$$I_{3R} = 0.3 \text{ A} \cdot \left( 1 - j \cdot 100.5 \right)$$

$$I_{3R} = 0.3 \text{ A} - j 30.15 \text{ A}$$

$$|I_{3R}| = \sqrt{0.3^2 + 30.15^2} = 30.16 \text{ A}$$

**Exercise E1 Complex Impedance Circuit**  
(written test, approx. 15 % of a 60-minute written test, WS2022)

**1. Calculate the current  $i(t)$  through the resistor  $R$  in the circuit shown in the figure. The voltage source is  $u(t) = 3.0 \text{ V} \cdot \sin(2\pi \cdot 15 \text{ kHz} \cdot t)$ .**

**Solution**

Result

$$Z = 48.2 \text{ } \Omega \quad Z_C = 19.8 \text{ } \Omega$$

Draw the circuit diagram of the given circuit

$$Z = \frac{U}{I} \quad I = \frac{U}{Z}$$

$$Z_C = \frac{1}{j\omega C} = \frac{1}{j \cdot 2\pi \cdot 15 \text{ kHz} \cdot 0.22 \text{ } \mu\text{F}}$$

$$Z = \sqrt{R^2 + (Z_L - Z_C)^2}$$

$$Z = \sqrt{10^2 + (100.5 - 19.8)^2} = 48.2 \text{ } \Omega$$

$$I = \frac{3.0 \text{ V}}{48.2 \text{ } \Omega} = 0.062 \text{ A}$$

$$i(t) = 0.062 \text{ A} \cdot \sin(2\pi \cdot 15 \text{ kHz} \cdot t)$$





**Exercise E7 Complex Impedance Circuit**  
**(written test, approx. 15 % of a 60-minute written test, WS2022)**

1. Calculate the current  $i(t)$  through the resistor  $R$  in the circuit shown in Fig. 1. The voltage source is  $u(t) = 3.0 \sin(2\pi \cdot 15 \text{ kHz} \cdot t)$  V. The circuit consists of a resistor  $R = 10 \text{ } \Omega$ , an inductor  $L = 330 \text{ } \mu\text{H}$ , and a capacitor  $C = 0.22 \text{ } \mu\text{F}$ , all in series.

Result:  $Z = 19.8 \text{ } \Omega$ ,  $i(t) = 0.152 \sin(2\pi \cdot 15 \text{ kHz} \cdot t - 90^\circ)$  A

Draw the circuit diagram of the given circuit and label all components, voltages, and currents.

$$Z = \frac{\hat{U}}{\hat{I}} \quad \hat{I} = \frac{\hat{U}}{Z} \quad Z_C = \frac{1}{j\omega C} = \frac{1}{j2\pi \cdot 15 \text{ kHz} \cdot 0.22 \text{ } \mu\text{F}}$$

Result:  $Z_C = -j19.2 \text{ } \Omega$

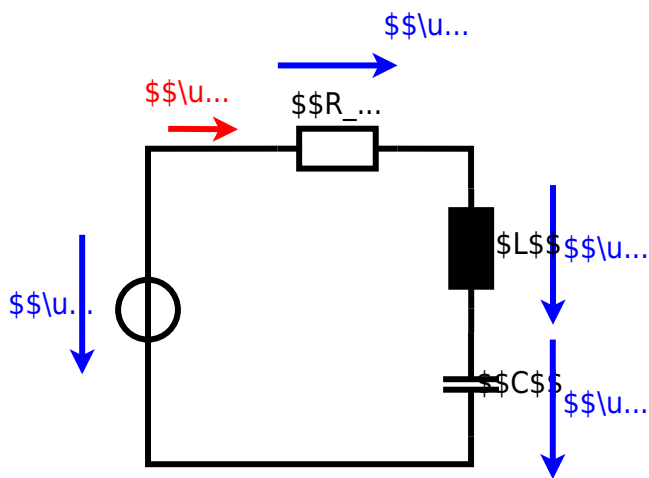
$$\hat{I} = \frac{\hat{U}}{Z} = \frac{3.0 \text{ V}}{19.8 \text{ } \Omega} = 0.152 \text{ A}$$

$$i(t) = 0.152 \sin(2\pi \cdot 15 \text{ kHz} \cdot t - 90^\circ) \text{ A}$$

$$\underline{Z} = R + \underline{Z}_L + \underline{Z}_C = R + j\omega L - j\omega C = R + j(\omega L - \omega C) = \sqrt{R^2 + (\omega L - \omega C)^2}$$







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